

PhD Qualifying Exam, August 2025

Opti 501, Day 1

System of units: SI (or MKSA)

- a) Write Maxwell's differential equations in their most complete form, including contributions from free-charge and free-current densities, as well as those from polarization and magnetization sources. Explain the meaning of each symbol that appears in these equations.
 - b) Derive the charge-current continuity equation directly from Maxwell's equations, and explain the meaning of this equation. Be brief but precise.
 - c) Define the bound-electric-charge and bound-electric-current densities. Use these entities to eliminate the $\mathbf{D}$ and $\mathbf{H}$ fields from Maxwell's equations. (In other words, rewrite Maxwell's equations with the help of bound-charge and bound-current densities in such a way that only the $\mathbf{E}$ and $\mathbf{B}$ fields would appear in the equations.)
 - d) Show that the bound-charge and bound-current densities of part (c) satisfy their own charge-current continuity equation.
-

Solution to Day 1 Problem)

$$\begin{aligned}
\text{a)} \quad \nabla \cdot \mathbf{D}(\mathbf{r}, t) &= \rho_{\text{free}}(\mathbf{r}, t), \\
\nabla \times \mathbf{H}(\mathbf{r}, t) &= \mathbf{J}_{\text{free}}(\mathbf{r}, t) + \partial \mathbf{D}(\mathbf{r}, t) / \partial t, \\
\nabla \times \mathbf{E}(\mathbf{r}, t) &= -\partial \mathbf{B}(\mathbf{r}, t) / \partial t, \\
\nabla \cdot \mathbf{B}(\mathbf{r}, t) &= 0.
\end{aligned}$$

In these equations, $\mathbf{r} = x\hat{\mathbf{x}} + y\hat{\mathbf{y}} + z\hat{\mathbf{z}}$ is an arbitrary point in space, while t is an arbitrary instant in time. $\mathbf{E}$ is the electric field, $\mathbf{H}$ is the magnetic field, $\mathbf{D}$ is the displacement, and $\mathbf{B}$ is the magnetic induction. The fields are related to each other, to the permittivity and permeability of free space, ϵ_0 and μ_0 , and to polarization $\mathbf{P}$ and magnetization $\mathbf{M}$, as follows:

$$\begin{aligned}
\mathbf{D}(\mathbf{r}, t) &= \epsilon_0 \mathbf{E}(\mathbf{r}, t) + \mathbf{P}(\mathbf{r}, t), \\
\mathbf{B}(\mathbf{r}, t) &= \mu_0 \mathbf{H}(\mathbf{r}, t) + \mathbf{M}(\mathbf{r}, t).
\end{aligned}$$

The sources of the electromagnetic fields (namely, $\mathbf{E}$ and $\mathbf{H}$) are the free charge-density ρ_{free} , free current-density $\mathbf{J}_{\text{free}}$, polarization $\mathbf{P}$ (which is the density of electric dipole moments), and magnetization $\mathbf{M}$ (the density of magnetic dipole moments). The operator $\partial/\partial t$ represents partial differentiation with respect to time, $\nabla \cdot$ is the divergence operator, and $\nabla \times$ is the curl operator. The divergence of a vector field such as $\mathbf{D}(\mathbf{r}, t)$, which turns out to be a scalar field, is defined as the integral of $\mathbf{D}(\mathbf{r}, t)$ over a small, closed surface, normalized by the enclosed volume. The curl of a vector field such as $\mathbf{E}(\mathbf{r}, t)$, which turns out to be another vector field, when projected onto the surface normal of a small surface element, yields the line integral of $\mathbf{E}(\mathbf{r}, t)$ around the boundary of the surface element, normalized by the elemental surface area.

b) To derive the charge-current continuity equation from Maxwell's equations, apply the divergence operator to both sides of the second (Maxwell-Ampere) equation. The divergence of curl is always equal to zero and, therefore, the left-hand-side of the equation becomes $\nabla \cdot (\nabla \times \mathbf{H}) = 0$. The right-hand side, $\nabla \cdot \mathbf{J}_{\text{free}} + \partial(\nabla \cdot \mathbf{D})/\partial t$, thus becomes zero. Maxwell's first equation (Gauss's law) now allows one to replace $\nabla \cdot \mathbf{D}$ with ρ_{free} , yielding the continuity equation as $\nabla \cdot \mathbf{J}_{\text{free}} + \partial \rho_{\text{free}} / \partial t = 0$. This equation informs that the integrated free current over any closed surface is precisely balanced by changes in the electrical charge contained within the closed surface. If there is a net outflow of current, the charge within the closed surface must be decreasing, and if there is a net inflow of current, the charge within must be increasing.

c) In the first of Maxwell's equations, we substitute $\mathbf{D} = \epsilon_0 \mathbf{E} + \mathbf{P}$ and obtain

$$\nabla \cdot (\epsilon_0 \mathbf{E} + \mathbf{P}) = \rho_{\text{free}} \rightarrow \epsilon_0 \nabla \cdot \mathbf{E} = \rho_{\text{free}} - \nabla \cdot \mathbf{P} \rightarrow \epsilon_0 \nabla \cdot \mathbf{E} = \rho_{\text{free}} + \rho_{\text{bound}}^{(e)}.$$

The bound-charge density is thus seen to be $\rho_{\text{bound}}^{(e)}(\mathbf{r}, t) = -\nabla \cdot \mathbf{P}(\mathbf{r}, t)$.

In the second Maxwell equation, we multiply both sides by μ_0 , then add $\nabla \times \mathbf{M}$ to both sides, in order to replace $\mathbf{H}$ with $\mathbf{B}$ through the identity $\mathbf{B} = \mu_0 \mathbf{H} + \mathbf{M}$. We also use $\mathbf{D} = \epsilon_0 \mathbf{E} + \mathbf{P}$ on the right-hand side of the equation to get rid of $\mathbf{D}$. We will have

$$\mu_0 \nabla \times \mathbf{H} + \nabla \times \mathbf{M} = \mu_0 \mathbf{J}_{\text{free}} + \mu_0 \frac{\partial(\epsilon_0 \mathbf{E} + \mathbf{P})}{\partial t} + \nabla \times \mathbf{M}$$

$$\rightarrow \quad \nabla \times \mathbf{B} = \mu_0 (\mathbf{J}_{\text{free}} + \partial \mathbf{P} / \partial t + \mu_0^{-1} \nabla \times \mathbf{M}) + \mu_0 \epsilon_0 \partial \mathbf{E} / \partial t$$

$$\rightarrow \quad \nabla \times \mathbf{B} = \mu_0 (\mathbf{J}_{\text{free}} + \mathbf{J}_{\text{bound}}^{(e)}) + \mu_0 \epsilon_0 \partial \mathbf{E} / \partial t.$$

The bound electric current-density is thus found to be $\mathbf{J}_{\text{bound}}^{(e)} = \partial \mathbf{P} / \partial t + \mu_0^{-1} \nabla \times \mathbf{M}$. Since the remaining Maxwell equations do not contain $\mathbf{D}$ and $\mathbf{H}$, they remain unchanged.

d) The divergence of $\mathbf{J}_{\text{bound}}^{(e)}$ is readily obtained, as follows:

$$\nabla \cdot \mathbf{J}_{\text{bound}}^{(e)} = \partial(\nabla \cdot \mathbf{P}) / \partial t + \mu_0^{-1} \nabla \cdot (\nabla \times \mathbf{M}).$$

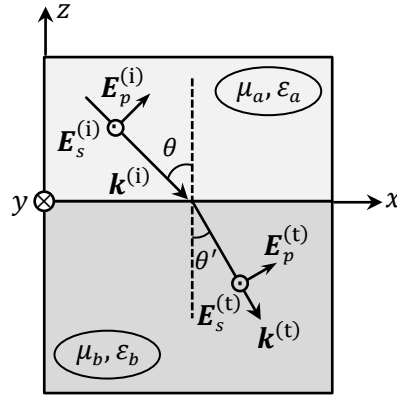
On the right-hand side of this equation, the divergence of the curl is always zero. Also the divergence of $\mathbf{P}(\mathbf{r}, t)$ is, by definition, $-\rho_{\text{bound}}^{(e)}$. Therefore, $\nabla \cdot \mathbf{J}_{\text{bound}}^{(e)} + \partial \rho_{\text{bound}}^{(e)} / \partial t = 0$. This is the charge-current continuity equation for the bound electrical charge and current defined in part (c).

PhD Qualifying Exam, August 2025

Opti 501, Day 2

System of units: SI (or MKSA)

Consider the flat interface between two linear, isotropic, homogeneous media specified by their relative permeability and permittivity at the incidence frequency, namely, (μ_a, ϵ_a) for the medium above, and (μ_b, ϵ_b) for the medium below the interface. These material parameters (i.e., $\mu_a, \mu_b, \epsilon_a, \epsilon_b$) are assumed to be real-valued and positive. A homogeneous plane-wave of frequency ω arrives at the interfacial xy -plane; the plane of incidence is xz , the incidence angle is θ , and the E -field components of the incident beam are $\mathbf{E}_p^{(i)}$ and $\mathbf{E}_s^{(i)}$, as indicated in the figure. The E -field components of the transmitted beam, also a homogeneous plane-wave, are $\mathbf{E}_p^{(t)}$ and $\mathbf{E}_s^{(t)}$, and the angle between the transmitted k -vector and the surface-normal is θ' , as shown.



- Invoking the dispersion relation $\mathbf{k} \cdot \mathbf{k} = (\omega/c)^2 \mu(\omega) \epsilon(\omega)$, write expressions for the k -vectors of the incident and transmitted plane-waves shown in the above figure. ($c = 1/\sqrt{\mu_0 \epsilon_0}$ is the speed of light in vacuum.)
 - Invoking Maxwell's boundary conditions, explain why the transmitted wave has the same frequency ω as the incident wave. What do these boundary conditions reveal about the relation between θ and θ' ?
 - Use Maxwell's third equation, $\nabla \times \mathbf{E} = -\partial \mathbf{B} / \partial t$, to determine both the incident and the transmitted H -field components. (As usual, $\mathbf{B} = \mu_0 \mu(\omega) \mathbf{H}$; you may use the impedance of free space, $Z_0 = \sqrt{\mu_0 / \epsilon_0}$, to simplify the equations.)
 - Find the conditions under which the reflected beam for the p -polarized incident light vanishes.
 - Find the conditions under which the reflected beam for the s -polarized incident light vanishes.
-

Solution to Day 2 Problem)

$$\text{a)} \quad |\mathbf{k}^{(i)}| = (\omega/c)\sqrt{\mu_a \varepsilon_a} \rightarrow \mathbf{k}^{(i)} = (\omega/c)\sqrt{\mu_a \varepsilon_a}(\sin \theta \hat{\mathbf{x}} - \cos \theta \hat{\mathbf{z}}). \quad (1)$$

$$|\mathbf{k}^{(t)}| = (\omega/c)\sqrt{\mu_b \varepsilon_b} \rightarrow \mathbf{k}^{(t)} = (\omega/c)\sqrt{\mu_b \varepsilon_b}(\sin \theta' \hat{\mathbf{x}} - \cos \theta' \hat{\mathbf{z}}). \quad (2)$$

b) Maxwell's boundary conditions require that $\mathbf{E}_{\parallel}$, $\mathbf{H}_{\parallel}$, $\mathbf{D}_{\perp}$, and $\mathbf{B}_{\perp}$ be continuous at the interface. Each field has a phase-factor $e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$, which reduces to $e^{i(k_x x + k_y y - \omega t)}$ when the interfacial plane is chosen to be the xy -plane at $z = 0$. Since the continuity conditions pertain to the fields immediately above and immediately below the interface at all times t , the frequencies of the incident, reflected, and transmitted beams must be identical. In particular, the frequency of the transmitted beam must be the same as the frequency ω of the incident beam.

Similarly, the continuity conditions are satisfied for all values of the coordinate y at the interfacial plane if and only if the k_y values of the incident, reflected, and transmitted beams are identical. Since our choice of xz as the plane of incidence automatically sets the k_y component of $\mathbf{k}^{(i)}$ to zero, we conclude that the k_y components of $\mathbf{k}^{(r)}$ and $\mathbf{k}^{(t)}$ must be zero as well.

Finally, the satisfaction of the boundary conditions for all values of the coordinate x at the interfacial plane requires that the k_x values of the incident, reflected, and transmitted beams be identical. In particular, setting $k_x^{(i)} = k_x^{(t)}$, we find from Eqs.(1) and (2) that the angles θ and θ' must be related as follows:

$$(\omega/c)\sqrt{\mu_a \varepsilon_a} \sin \theta = (\omega/c)\sqrt{\mu_b \varepsilon_b} \sin \theta' \rightarrow \sin \theta' = \sqrt{(\mu_a \varepsilon_a)/(\mu_b \varepsilon_b)} \sin \theta. \quad (3)$$

c) From $\nabla \times \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} = -(\partial/\partial t)[\mu_0 \mu(\omega) \mathbf{H}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}]$ we find $\mathbf{k} \times \mathbf{E}_0 = \mu_0 \mu(\omega) \omega \mathbf{H}_0$, which leads to $(\omega/c)\sqrt{\mu(\omega)\varepsilon(\omega)}\hat{\mathbf{k}} \times \mathbf{E}_0 = \mu_0 \mu(\omega) \omega \mathbf{H}_0$ and, therefore, $\mathbf{H}_0 = \sqrt{\varepsilon/\mu} \hat{\mathbf{k}} \times \mathbf{E}_0 / Z_0$. For the incident plane-wave, this equation yields

$$\begin{aligned} \mathbf{H}_0^{(i)} &= \sqrt{\varepsilon_a/\mu_a} \hat{\mathbf{k}}^{(i)} \times \mathbf{E}_0^{(i)} / Z_0 \\ &= Z_0^{-1} \sqrt{\varepsilon_a/\mu_a} (\sin \theta \hat{\mathbf{x}} - \cos \theta \hat{\mathbf{z}}) \times [E_p^{(i)} \cos \theta \hat{\mathbf{x}} + E_s^{(i)} \hat{\mathbf{y}} + E_p^{(i)} \sin \theta \hat{\mathbf{z}}] \\ &= Z_0^{-1} \sqrt{\varepsilon_a/\mu_a} [E_s^{(i)} \cos \theta \hat{\mathbf{x}} - E_p^{(i)} (\sin^2 \theta + \cos^2 \theta) \hat{\mathbf{y}} + E_s^{(i)} \sin \theta \hat{\mathbf{z}}] \\ &= Z_0^{-1} \sqrt{\varepsilon_a/\mu_a} [E_s^{(i)} \cos \theta \hat{\mathbf{x}} - E_p^{(i)} \hat{\mathbf{y}} + E_s^{(i)} \sin \theta \hat{\mathbf{z}}]. \end{aligned} \quad (4)$$

Similarly, for the transmitted plane-wave, we will have

$$\mathbf{H}_0^{(t)} = \sqrt{\varepsilon_b/\mu_b} \hat{\mathbf{k}}^{(t)} \times \mathbf{E}_0^{(t)} / Z_0 = Z_0^{-1} \sqrt{\varepsilon_b/\mu_b} [E_s^{(t)} \cos \theta' \hat{\mathbf{x}} - E_p^{(t)} \hat{\mathbf{y}} + E_s^{(t)} \sin \theta' \hat{\mathbf{z}}]. \quad (5)$$

d) In the absence of a reflected beam, the continuity conditions for $\mathbf{E}_{\parallel}$ and $\mathbf{H}_{\parallel}$ of p -polarized light become

$$E_x^{(i)} = E_x^{(t)} \rightarrow E_p^{(i)} \cos \theta = E_p^{(t)} \cos \theta'. \quad (6)$$

$$\boxed{\text{See Eqs.(4) and (5)}} \rightarrow H_y^{(i)} = H_y^{(t)} \rightarrow \sqrt{\varepsilon_a/\mu_a} E_p^{(i)} = \sqrt{\varepsilon_b/\mu_b} E_p^{(t)}. \quad (7)$$

Substituting for $E_p^{(t)}$ from Eq.(7) into Eq.(6), and recalling the relation between θ and θ' as given by Eq.(3), we find

$$\begin{aligned} \cancel{E_p^{(i)}} \sqrt{1 - \sin^2 \theta} &= \sqrt{\mu_b \varepsilon_a / \mu_a \varepsilon_b} \cancel{E_p^{(i)}} \sqrt{1 - \sin^2 \theta'} \\ \rightarrow 1 - \sin^2 \theta &= (\mu_b \varepsilon_a / \mu_a \varepsilon_b) [1 - (\mu_a \varepsilon_a / \mu_b \varepsilon_b) \sin^2 \theta] \rightarrow \sin \theta = \sqrt{\frac{1 - (\mu_b \varepsilon_a / \mu_a \varepsilon_b)}{1 - (\varepsilon_a / \varepsilon_b)^2}}. \end{aligned} \quad (8)$$

If $\mu_a = \mu_b$, we will have $\sin \theta = \sqrt{\varepsilon_b / (\varepsilon_a + \varepsilon_b)}$, which leads to $\cos \theta = \sqrt{\varepsilon_a / (\varepsilon_a + \varepsilon_b)}$ and $\tan \theta = \sqrt{\varepsilon_b / \varepsilon_a}$. But this may also be written as $\tan \theta = \sqrt{\mu_b \varepsilon_b / \mu_a \varepsilon_a} = n_b / n_a$, which is the well-known result associated with p -light incidence at Brewster's angle when $\mu_a = \mu_b$.

e) In the case of an s -polarized incident beam, the reflected beam vanishes when the following continuity conditions for $\mathbf{E}_{\parallel}$ and $\mathbf{H}_{\parallel}$ are satisfied:

$$E_y^{(i)} = E_y^{(t)} \rightarrow E_s^{(i)} = E_s^{(t)}. \quad (9)$$

$$\text{See Eqs.(4) and (5)} \rightarrow H_x^{(i)} = H_x^{(t)} \rightarrow \sqrt{\varepsilon_a / \mu_a} E_s^{(i)} \cos \theta = \sqrt{\varepsilon_b / \mu_b} E_s^{(t)} \cos \theta'. \quad (10)$$

Substituting for $E_s^{(t)}$ from Eq.(9) into Eq.(10), and recalling the relation between θ and θ' as given by Eq.(3), we find

$$\begin{aligned} (\varepsilon_a / \mu_a) (1 - \sin^2 \theta) &= (\varepsilon_b / \mu_b) (1 - \sin^2 \theta') \\ \rightarrow (\mu_b \varepsilon_a / \mu_a \varepsilon_b) (1 - \sin^2 \theta) &= 1 - (\mu_a \varepsilon_a / \mu_b \varepsilon_b) \sin^2 \theta \rightarrow \sin \theta = \sqrt{\frac{\mu_b (\mu_a \varepsilon_b - \mu_b \varepsilon_a)}{(\mu_a^2 - \mu_b^2) \varepsilon_a}}. \end{aligned} \quad (11)$$

At optical frequencies, ordinary materials have $\mu_a = \mu_b \cong 1$, which does not allow for the existence of a Brewster's angle for s -polarized light. However, whenever $\mu_a \neq \mu_b$, if Eq.(11) yields an acceptable value for the angle θ (i.e., an angle in the range of 0° to 90°), then such a Brewster angle for s -light would exist. If it so happens that $\varepsilon_a = \varepsilon_b$ while $\mu_a \neq \mu_b$, we will have $\sin \theta = \sqrt{\mu_b / (\mu_a + \mu_b)}$, which leads to $\cos \theta = \sqrt{\mu_a / (\mu_a + \mu_b)}$ and $\tan \theta = \sqrt{\mu_b / \mu_a}$. Once again, this may be written as $\tan \theta = \sqrt{\mu_b \varepsilon_b / \mu_a \varepsilon_a} = n_b / n_a$, as was the case for p -polarized light.
