

Problem 40)

a) Electric dipole:

$$\vec{E}(\vec{r}, t) = \frac{Z_0 I_0 d}{4\pi} \left\{ \left[\frac{1}{r^2} \sin(\omega t - k_0 r) - \frac{(\lambda_0 / 2\pi)}{r^3} \cos(\omega t - k_0 r) \right] (2 \cos \theta \hat{r} + \sin \theta \hat{\theta}) + \frac{k_0}{r} \sin \theta \cos(\omega t - k_0 r) \hat{\theta} \right\}$$

$$\vec{H}(\vec{r}, t) = \frac{I_0 d}{4\pi} \sin \theta \left\{ \frac{1}{r^2} \sin(\omega t - k_0 r) + \frac{k_0}{r} \cos(\omega t - k_0 r) \right\} \hat{\phi}$$

$$\vec{S}(\vec{r}, t) = \vec{E}(\vec{r}, t) \times \vec{H}(\vec{r}, t) = Z_0 \left(\frac{I_0 d}{4\pi} \right)^2 \sin \theta \left\{ \left[\frac{\sin(\omega t - k_0 r)}{r^2} - \frac{\cos(\omega t - k_0 r)}{k_0 r^3} \right] \right.$$

$$\left. \left[\frac{\sin(\omega t - k_0 r)}{r^2} + \frac{k_0 \cos(\omega t - k_0 r)}{r} \right] (-2 \cos \theta \hat{\theta} + \sin \theta \hat{r}) + \frac{k_0}{r} \sin \theta \cos(\omega t - k_0 r) \left[\frac{\sin(\omega t - k_0 r)}{r^2} + \frac{k_0 \cos(\omega t - k_0 r)}{r} \right] \hat{r} \right\} \Rightarrow$$

$$\vec{S}(\vec{r}, t) = Z_0 \left(\frac{I_0 d}{4\pi} \right)^2 \sin \theta \left\{ \left[\frac{\sin^2(\omega t - k_0 r)}{r^4} - \frac{\cos^2(\omega t - k_0 r)}{r^4} + \frac{1}{2} \sin 2(\omega t - k_0 r) \left(\frac{k_0}{r^3} - \frac{1}{k_0 r^5} \right) \right] \right.$$

$$\left. (\sin \theta \hat{r} - 2 \cos \theta \hat{\theta}) + \left[\frac{k_0 \sin \theta}{2 r^3} \sin 2(\omega t - k_0 r) + \sin \theta \left(\frac{k_0}{r} \right)^2 \cos^2(\omega t - k_0 r) \right] \hat{r} \right\} \Rightarrow$$

$$\vec{S}(\vec{r}, t) = Z_0 \left(\frac{I_0 d}{4\pi} \right)^2 \left\{ \left[\left(\frac{k_0}{2 r^3} - \frac{1}{2 k_0 r^5} \right) \sin(2\omega t - 2k_0 r) - \frac{1}{r^4} \cos(2\omega t - 2k_0 r) \right] (\sin^2 \theta \hat{r} - \sin 2\theta \hat{\theta}) + \left[\frac{k_0 \sin^2 \theta}{2 r^3} \sin(2\omega t - 2k_0 r) + \left(\frac{k_0 \sin \theta}{r} \right)^2 \cos^2(\omega t - k_0 r) \right] \hat{r} \right\}$$

Here $k_0 = 2\pi/\lambda_0 = 2\pi f/c = \omega/c$ and $f = 1/T$, where T is the period of oscillations.

The time-averaged pointing vector is the average of $\vec{S}(\vec{r}, t)$ over one period of oscillations; that is,

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{1}{T} \int_0^T \vec{S}(\vec{r}, t) dt.$$

Since the time-averages of $\sin(2\omega t - 2k_0 r)$ and $\cos(2\omega t - 2k_0 r)$ are zero, and the time-average of $\cos^2(\omega t - k_0 r) = \frac{1}{2} + \frac{1}{2} \cos(2\omega t - 2k_0 r)$ is equal to $1/2$, we'll have:

$$\langle \vec{S}(\vec{r}, t) \rangle = \epsilon_0 \left(\frac{I_0 d}{4\pi} \right)^2 \left(\frac{k_0 \sin \theta}{r} \right)^2 \langle \cos^2(\omega t - k_0 r) \rangle \hat{r} \Rightarrow$$

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{1}{8} \epsilon_0 I_0^2 (d/\lambda_0)^2 \frac{\sin^2 \theta}{r^2} \hat{r}$$

If the total radiated power is desired, the above time-averaged Poynting vector may be integrated over the surface of a sphere of fixed radius r . The result is:

$$\text{Total radiated power} = \int_{\theta=0}^{\pi} 2\pi r^2 \sin \theta \langle S(\vec{r}, t) \rangle d\theta = \frac{\pi \epsilon_0}{4} (I_0 d/\lambda_0)^2 \int_0^{\pi} \sin^3 \theta d\theta$$

$$\text{where } \int_0^{\pi} \sin^3 \theta d\theta = \int_0^{\pi} \sin \theta (1 - \cos^2 \theta) d\theta = \int_0^{\pi} \sin \theta d\theta - \int_0^{\pi} \sin \theta \cos^2 \theta d\theta = -\cos \theta \Big|_0^{\pi} + \frac{1}{3} \cos^3 \theta \Big|_0^{\pi} = 2 - \frac{2}{3} = \frac{4}{3}. \text{ Therefore,}$$

$$\text{Total radiated power} = \frac{1}{3} \pi \epsilon_0 (I_0 d/\lambda_0)^2$$

Since the radiation fields for a dipole that we have obtained are not valid in the immediate vicinity of the dipole itself (i.e., in the near near field) we cannot calculate the \vec{E} -field that must be applied to the short dipole antenna in order to drive the current $I_0 \sin(\omega t)$ within the dipole.

b) Small loop of current (magnetic dipole): We saw in HW #6, Prob. 4:

$$\vec{E}(\vec{r}, t) = \frac{\epsilon_0 m_0 k_0}{4\pi} \sin \theta \left\{ \frac{1}{r^2} \sin(\omega t - k_0 r) + \frac{k_0}{r} \cos(\omega t - k_0 r) \right\} \hat{\Phi}$$

$$\vec{H}(\vec{r}, t) = -\frac{m_0 k_0}{4\pi} \left\{ \left[\frac{1}{r^2} \sin(\omega t - k_0 r) - \frac{(\lambda_0/2\pi)}{r^3} \cos(\omega t - k_0 r) \right] (2\cos \theta \hat{r} + \sin \theta \hat{\theta}) + \frac{k_0}{r} \sin \theta \cos(\omega t - k_0 r) \hat{\Phi} \right\}$$

These fields are identical with those of an electric dipole studied in part (a) if the roles of \vec{E} and \vec{H} are reversed, and if $m_0 k_0$ is substituted for $I_0 d$. Note

that the units of $m_0 k_0$ and $I_0 d$ are the same, namely, Amp-meter.

Again the expressions for \vec{E} and \vec{H} are not valid in the near-field, so we cannot use them to determine the necessary voltage to drive the current around the loop.

C) Hollow Cylinder (Radius = R): We saw in HW#7, Problem 2:

$$\vec{E}(\vec{r}, t) = \frac{1}{4} \epsilon_0 I_0 k_0 \begin{cases} J_0(k_0 R) [Y_0(k_0 \rho) \cos \omega t - J_0(k_0 \rho) \sin \omega t] \hat{z}; & \rho \geq R \\ [Y_0(k_0 R) \cos \omega t - J_0(k_0 R) \sin \omega t] J_0(k_0 \rho) \hat{z}; & \rho \leq R \end{cases}$$

$$\vec{H}(\vec{r}, t) = -\frac{1}{4} I_0 k_0 \begin{cases} J_0(k_0 R) [Y_1(k_0 \rho) \sin \omega t + J_1(k_0 \rho) \cos \omega t] \hat{\phi}; & \rho \geq R \\ [Y_0(k_0 R) \sin \omega t + J_0(k_0 R) \cos \omega t] J_1(k_0 \rho) \hat{\phi}; & \rho \leq R \end{cases}$$

First we compute the Poynting vector inside the cylinder ($\rho < R$):

$$\vec{S}(\rho, t) = \vec{E} \times \vec{H} = \frac{1}{16} \epsilon_0 (I_0 k_0)^2 \left[Y_0^2(k_0 R) \sin \omega t \cos \omega t - J_0^2(k_0 R) \sin \omega t \cos \omega t + Y_0(k_0 R) J_0(k_0 R) \cos^2 \omega t - Y_0(k_0 R) J_0(k_0 R) \sin^2 \omega t \right] J_0(k_0 \rho) J_1(k_0 \rho) \hat{\rho} \Rightarrow$$

$$\vec{S}(\rho, t) = \frac{\epsilon_0}{32} (I_0 k_0)^2 \left\{ [Y_0^2(k_0 R) - J_0^2(k_0 R)] \sin(2\omega t) + 2 Y_0(k_0 R) J_0(k_0 R) \cos(2\omega t) \right\} \times J_0(k_0 \rho) J_1(k_0 \rho) \hat{\rho}; \quad \rho < R$$

Since the time averages of $\sin(2\omega t)$ and $\cos(2\omega t)$ are zero, we'll have:

$$\langle \vec{S}(\rho, t) \rangle = 0; \quad \rho < R$$

The Poynting vector outside the cylinder is given by:

$$\vec{S}(p, t) = \frac{1}{16} \epsilon_0 (I_0 k_0)^2 J_0^2(k_0 R) \left\{ \frac{1}{2} [\gamma_0(k_0 p) \gamma_1(k_0 p) - J_0(k_0 p) J_1(k_0 p)] \sin(2\omega t) + \gamma_0(k_0 p) J_1(k_0 p) \cos^2 \omega t - J_0(k_0 p) \gamma_1(k_0 p) \sin^2 \omega t \right\} \hat{p}; \quad p \geq R$$

$$\langle \vec{S}(p, t) \rangle = \frac{1}{32} \epsilon_0 (I_0 k_0)^2 J_0^2(k_0 R) \left[\gamma_0(k_0 p) J_1(k_0 p) - J_0(k_0 p) \gamma_1(k_0 p) \right] \hat{p} \Rightarrow$$

$\rightarrow = 2/\pi k_0 p$; see the note on Bessel functions.

$$\langle \vec{S}(p, t) \rangle = \frac{\epsilon_0}{16\pi k_0 p} [I_0 k_0 J_0(k_0 R)]^2 \hat{p} = \frac{\epsilon_0 I_0^2 k_0}{16\pi p} J_0^2(k_0 R) \hat{p}; \quad p \geq R$$

The radiated power that leaves a cylinder of radius p and unit length is thus found to be:

$$\text{Total radiated power per unit length} = 2\pi p \langle S \rangle = \frac{1}{8} \epsilon_0 I_0^2 k_0 J_0^2(k_0 R)$$

Now, the \vec{E} -field at the cylinder surface is obtained by setting $p=R$ in the expression for the \vec{E} -field:

$$\vec{E}(\vec{r}, t) = \frac{1}{4} \epsilon_0 I_0 k_0 J_0(k_0 R) [\gamma_0(k_0 R) \cos \omega t - J_0(k_0 R) \sin \omega t] \hat{z}$$

This \vec{E} -field opposes the \vec{E} -field of the external source that drives the current within the walls of the cylindrical conductor. In a perfect conductor, the net \vec{E} -field must be zero; therefore, the external field is exactly equal to the negative of the above, namely,

$$\vec{E}_{\text{external}} = -\frac{1}{4} \epsilon_0 I_0 k_0 J_0(k_0 R) [\gamma_0(k_0 R) \cos \omega t - J_0(k_0 R) \sin \omega t] \hat{z}$$

The ^{electric} power per unit length of the cylinder supplied by the external source is:

$$\begin{aligned} \text{Electrical Power per unit length} &= E_{\text{external}} I = E_{\text{external}} I_0 \sin \omega t = \\ &= -\frac{1}{4} \epsilon_0 I_0^2 k_0 J_0(k_0 R) [\gamma_0(k_0 R) \cos \omega t \sin \omega t - J_0(k_0 R) \sin^2 \omega t] \\ &= \frac{1}{8} \epsilon_0 I_0^2 k_0 J_0(k_0 R) [J_0(k_0 R) - J_0(k_0 R) \cos(2\omega t) + \gamma_0(k_0 R) \sin(2\omega t)] \end{aligned}$$

Time averaged electrical power per unit length = $\frac{1}{2} \epsilon_0 I_0^2 k_0 J_0^2 (k_0 R)$
 is thus in complete agreement with the expression obtained earlier
 for the time-averaged radiated power per unit length of the cylinder.

d) Infinite sheet of current in the yz -plane; $\vec{J}_s(t) = J_{s0} \sin(\omega t) \hat{z}$; HW#7, Prob. 3:

$$\begin{cases} \vec{E}(\vec{r}, t) = \frac{1}{2} \epsilon_0 J_{s0} \sin(k_0 x - \omega t) \hat{z}; & x > 0 \\ \vec{H}(\vec{r}, t) = -\frac{1}{2} J_{s0} \sin(k_0 x - \omega t) \hat{y}; & x > 0 \end{cases}$$

$$\vec{S}(\vec{r}, t) = \vec{E} \times \vec{H} = \frac{1}{4} \epsilon_0 J_{s0}^2 \sin^2(k_0 x - \omega t) \hat{x}; \quad x > 0$$

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{1}{8} \epsilon_0 J_{s0}^2 \hat{x}; \quad x > 0 \quad \leftarrow \text{Radiated power per unit area || to } xz\text{-plane.}$$

\vec{E} -field at the surface of the conductor is obtained by setting $x=0$ in the expression for the \vec{E} -field:

$$\vec{E}(x=0, t) = -\frac{1}{2} \epsilon_0 J_{s0} \sin(\omega t) \hat{z} \quad (\text{in the } yz\text{-plane})$$

$$\vec{E}_{\text{external}} = -\vec{E}(x=0, t) = \frac{1}{2} \epsilon_0 J_{s0} \sin(\omega t) \hat{z}$$

Electrical power supplied to the sheet (per unit area) = $\vec{E}_{\text{external}} \cdot \vec{J}_s =$

$$\left[\frac{1}{2} \epsilon_0 J_{s0} \sin(\omega t) \right] J_{s0} \sin(\omega t) = \frac{1}{2} \epsilon_0 J_{s0}^2 \sin^2(\omega t)$$

The time-averaged electrical power (per unit area of the sheet) is thus $\frac{1}{4} \epsilon_0 J_{s0}^2$, which is twice the radiated power in the $x > 0$ region.

Considering that the same amount of optical power is radiated in the $x < 0$ region, there is complete agreement between the total electrical power supplied and the total optical power radiated.

e) Infinite Sheet of Current $\vec{J}_s(\vec{r}, t) = J_{s0} \sin(\omega t - \kappa z) \hat{z}$ located in yz -plane at $x=0$.

Case I: $\kappa < k_0$ From HW#7 Prob. 4 and also from the discussion in the class we have:

$$\left\{ \begin{aligned} \vec{E}(\vec{r}, t) &= -\frac{1}{2} \epsilon_0 J_{s0} \left(\frac{\kappa}{k_0} \hat{x} - \sqrt{1 - (\kappa/k_0)^2} \hat{z} \right) \sin(\sqrt{k_0^2 - \kappa^2} x + \kappa z - \omega t); & x > 0 \\ H(\vec{r}, t) &= -\frac{1}{2} J_{s0} \sin(\sqrt{k_0^2 - \kappa^2} x + \kappa z - \omega t) \hat{y}; & x > 0 \end{aligned} \right.$$

$$\vec{S}(\vec{r}, t) = \vec{E} \times \vec{H} = \frac{1}{4} \epsilon_0 J_{s0}^2 \left(\frac{\kappa}{k_0} \hat{z} + \sqrt{1 - (\kappa/k_0)^2} \hat{x} \right) \sin^2 \left[k_0 \left(\sqrt{1 - (\kappa/k_0)^2} x + \frac{\kappa}{k_0} z - ct \right) \right]$$

We define a unit vector $\vec{\sigma} = \sqrt{1 - (\kappa/k_0)^2} \hat{x} + (\kappa/k_0) \hat{z}$ to represent the direction of propagation of energy. We'll have:

$$\vec{S}(\vec{r}, t) = \frac{1}{4} \epsilon_0 J_{s0}^2 \sin^2 [k_0 (\vec{\sigma} \cdot \vec{r} - ct)] \vec{\sigma}; \quad x > 0$$

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{1}{8} \epsilon_0 J_{s0}^2 \vec{\sigma}; \quad x > 0$$

$$\vec{E}\text{-field acting on the current sheet} = \vec{E}(x=0, y, z, t) = -\frac{1}{2} \epsilon_0 J_{s0} \left(\frac{\kappa}{k_0} \hat{x} - \sqrt{1 - (\kappa/k_0)^2} \hat{z} \right) \sin(\kappa z - \omega t)$$

The only component of \vec{E} that does work on the current is E_z ; the external field applied by the power supply to counter this component of the radiated field is, therefore, given by:

$$\vec{E}_{\text{external}} = -E_z(x=0, y, z, t) \hat{z} = +\frac{1}{2} \epsilon_0 J_{s0} \sqrt{1 - (\kappa/k_0)^2} \sin(\omega t - \kappa z) \hat{z}$$

Electrical power per unit area supplied by the external source =

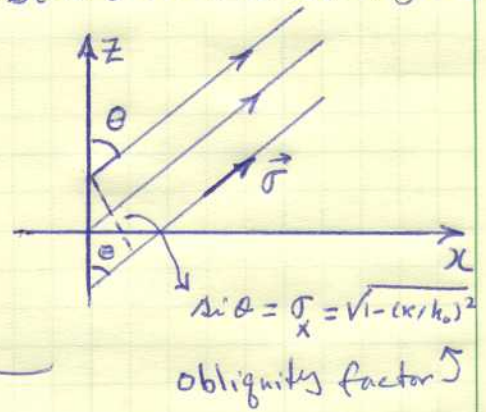
$$\vec{E}_{\text{external}} \cdot \vec{J}_s = \frac{1}{2} \epsilon_0 J_{s0}^2 \sqrt{1 - (\kappa/k_0)^2} \sin^2(\omega t - \kappa z)$$

$$\text{Time-averaged electrical power per unit area of the sheet} = \frac{1}{4} \epsilon_0 J_{s0}^2 \sqrt{1 - (\kappa/k_0)^2}$$

Again, the factor of 2 difference between electrical and optical powers

is due to the fact that the sheet radiates into both $x > 0$ and $x < 0$ regions.

The factor $\sqrt{1 - (\kappa/k_0)^2}$ is the obliquity factor, accounting for the reduced cross-section of the beam when projected onto the plane perpendicular to the propagation direction $\vec{\sigma}$.



Case II: $\kappa > k_0$

$$\begin{cases} \vec{E}(\vec{r}, t) = -\frac{1}{2} \epsilon_0 J_{s0} e^{-\sqrt{\kappa^2 - k_0^2} x} \left[\frac{\kappa}{k_0} \sin(\kappa z - \omega t) \hat{x} - \sqrt{(\kappa/k_0)^2 - 1} \cos(\kappa z - \omega t) \hat{z} \right]; & x > 0 \\ \vec{H}(\vec{r}, t) = -\frac{1}{2} J_{s0} e^{-\sqrt{\kappa^2 - k_0^2} x} \sin(\kappa z - \omega t) \hat{y}; & x > 0 \end{cases}$$

$$\vec{S}(\vec{r}, t) = \vec{E} \times \vec{H} = \frac{1}{4} \epsilon_0 J_{s0}^2 e^{-2\sqrt{\kappa^2 - k_0^2} x} \left\{ \frac{\kappa}{k_0} \sin^2(\kappa z - \omega t) \hat{z} + \frac{1}{2} \sqrt{(\kappa/k_0)^2 - 1} \sin[2(\kappa z - \omega t)] \hat{x} \right\}$$

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{\kappa \epsilon_0 J_{s0}^2}{8k_0} \exp(-2\sqrt{\kappa^2 - k_0^2} x) \hat{z}$$

Note that in this case, when $\kappa > k_0$, the optical energy does not leave the vicinity of the current sheet; rather it lingers within a certain range of the x -axis, up to a distance of $\sim \lambda_0 / \sqrt{(\kappa/k_0)^2 - 1}$, and propagates along the z -axis.

$$E\text{-field acting on the current sheet} = -E_z(x=0, y, z, t) = -\frac{1}{2} \epsilon_0 J_{s0} \sqrt{(\kappa/k_0)^2 - 1} \cos(\kappa z - \omega t)$$

$$\begin{aligned} \text{Electrical Power (per unit area) delivered to the sheet} &= \vec{E}_{\text{external}} \cdot \vec{J}_s \\ &= \frac{1}{4} \epsilon_0 J_{s0}^2 \sqrt{(\kappa/k_0)^2 - 1} \sin[2(\kappa z - \omega t)] \end{aligned}$$

\Rightarrow Time-averaged electrical power per unit area = 0.

This shows that the optical energy in the system is recycled, i.e., what goes out at $z = +\infty$ returns at $z = -\infty$. Therefore, since no energy leaves the system in the steady-state, there is no need for external energy to be supplied.

f) Infinite sheet of current $\vec{J}_s(\vec{r}, t) = J_{s0} \sin(\omega t - \kappa y) \hat{z}$ located in the yz -plane at $x=0$.

Case I: $\kappa < k_0$ From HW #7, Problem 5, we have:

$$\vec{E}(\vec{r}, t) = -\frac{Z_0 J_{s0}}{2\sqrt{1-(\kappa/k_0)^2}} \sin(\omega t - \kappa y - \sqrt{k_0^2 - \kappa^2} x) \hat{z}; \quad x \geq 0$$

$$\vec{H}(\vec{r}, t) = -\frac{J_{s0}}{2\sqrt{1-(\kappa/k_0)^2}} \left(\frac{\kappa}{k_0} \hat{x} - \sqrt{1-(\kappa/k_0)^2} \hat{y} \right) \sin(\omega t - \kappa y - \sqrt{k_0^2 - \kappa^2} x); \quad x \geq 0$$

$$\vec{S}(\vec{r}, t) = \vec{E} \times \vec{H} = \frac{Z_0 J_{s0}^2}{4[1-(\kappa/k_0)^2]} \left(\frac{\kappa}{k_0} \hat{y} + \sqrt{1-(\kappa/k_0)^2} \hat{x} \right) \sin^2 \left[k_0 \left(\frac{\kappa}{k_0} y + \sqrt{1-(\kappa/k_0)^2} x - ct \right) \right]$$

Defining the unit vector $\vec{\sigma} = \sqrt{1-(\kappa/k_0)^2} \hat{x} + (\kappa/k_0) \hat{y}$ to represent the direction of propagation of energy, we write:

$$\vec{S}(\vec{r}, t) = \frac{Z_0 J_{s0}^2}{4[1-(\kappa/k_0)^2]} \sin^2 [k_0 (\vec{\sigma} \cdot \vec{r} - ct)] \vec{\sigma}; \quad x \geq 0$$

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{Z_0 J_{s0}^2}{8[1-(\kappa/k_0)^2]} \vec{\sigma}; \quad x \geq 0$$

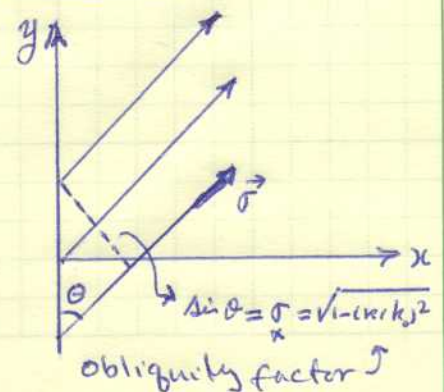
$$\vec{E}_{\text{external}} \text{ acting on the sheet} = -\vec{E}_z(x=0, y, z, t) \hat{z} = \frac{Z_0 J_{s0}}{2\sqrt{1-(\kappa/k_0)^2}} \sin(\omega t - \kappa y) \hat{z}$$

External electrical power per unit area delivered to the sheet by the external source =

$$\vec{E}_{\text{external}} \cdot \vec{J}_s = \frac{Z_0 J_{s0}^2}{2\sqrt{1-(\kappa/k_0)^2}} \sin^2(\omega t - \kappa y)$$

$$\text{Time-averaged electrical power per unit area} = \frac{Z_0 J_{s0}^2}{4\sqrt{1-(\kappa/k_0)^2}}$$

As before, the factor of 2 difference between optical and electrical powers can be accounted for by the radiation of the sheet into both $x \geq 0$ and $x \leq 0$ regions, and the factor $\sqrt{1-(\kappa/k_0)^2}$ is the obliquity factor.



Case II: $\kappa > k_0$

$$\vec{E}(\vec{r}, t) = -\frac{Z_0 J_{s0}}{2\sqrt{(\kappa/k_0)^2 - 1}} \exp(-\sqrt{\kappa^2 - k_0^2} x) \cos(\omega t - \kappa y) \hat{z}; \quad x \geq 0$$

$$\vec{H}(\vec{r}, t) = -\frac{J_{s0}}{2\sqrt{(\kappa/k_0)^2 - 1}} \left[\frac{\kappa}{k_0} \cos(\omega t - \kappa y) \hat{x} - \sqrt{(\kappa/k_0)^2 - 1} \sin(\omega t - \kappa y) \hat{y} \right] e^{-\sqrt{\kappa^2 - k_0^2} x}; \quad x \geq 0$$

$$\vec{S}(\vec{r}, t) = \vec{E} \times \vec{H} = \frac{Z_0 J_{s0}^2}{4[(\kappa/k_0)^2 - 1]} \exp(-2\sqrt{\kappa^2 - k_0^2} x) \left\{ \frac{\kappa}{k_0} \cos^2(\omega t - \kappa y) \hat{y} + \frac{1}{2} \sqrt{(\kappa/k_0)^2 - 1} \sin[2(\omega t - \kappa y)] \hat{x} \right\}$$

$$\langle \vec{S}(\vec{r}, t) \rangle = \frac{(\kappa/k_0) Z_0 J_{s0}^2}{8[(\kappa/k_0)^2 - 1]} \exp(-2k_0 \sqrt{(\kappa/k_0)^2 - 1} x) \hat{y}; \quad x \geq 0$$

$$\vec{E}_{\text{external}} = -E(x=0, y, z, t) \hat{z} = \frac{Z_0 J_{s0}}{2\sqrt{(\kappa/k_0)^2 - 1}} \cos(\omega t - \kappa y) \hat{z}$$

Electrical Power per unit area supplied to the current sheet by the external source = $\vec{E}_{\text{external}} \cdot \vec{J}_s = \frac{Z_0 J_{s0}^2}{4\sqrt{(\kappa/k_0)^2 - 1}} \sin[2(\omega t - \kappa y)]$

Time-averaged electrical power per unit area = 0.

As before, the optical energy in this case is recycled; what goes out at $y = \infty$ returns from $y = -\infty$. Since no energy leaves the system in the steady-state, there is no need for external energy to be supplied. All the optical energy in the system is confined to the vicinity of the current sheet, within a distance of $\sim \lambda_0 / \sqrt{(\kappa/k_0)^2 - 1}$ from the yz -plane. The optical energy does propagate, however, parallel to the current sheet and in the \hat{y} -direction.